

UNITED STATES PATENT APPLICATION

APPARATUS FOR ACQUIRING AND TRANSMITTING NEURAL SIGNALS
AND RELATED METHODS

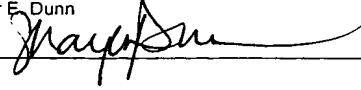
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Description

APPARATUS FOR ACQUIRING AND TRANSMITTING NEURAL SIGNALS
AND RELATED METHODS

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Grant Statement

This invention was supported by DARPA grant N0014-98-1-0676. Thus,
the Government has certain rights in this invention.

Technical Field

10

The present invention relates generally to apparatus for acquiring and
transmitting neural signals for processing. Specifically, the present invention
relates to a system for acquiring neural signals having neural spikes and
transmitting an information signal in response to detecting a neural spike.

15

Background Art

The human brain is an exceedingly complex processing system, which
integrates continual streams of incoming sensory input data with stored
memories, uses the input data and memories in complex decision processes at
both conscious and unconscious levels and, on the basis of these processes,
20 generates observable behaviors by activation of its motor or movement control
pathways and the muscles which these innervate. The neurons of the nervous

system propagate input data by generating characteristic electrical pulses called action potentials, or neural spikes, that can travel along nerve fibers. A single neuron or a group of neurons represent and transmit information by firing sequences of neural spikes in various temporal patterns. Information is carried
5 in the neural spike arrival times.

In certain cases of traumatic injury or neurological disease, the brain can be partially isolated from the periphery. Input data from certain senses are thus lost, at least for a portion of the body, as are many voluntary movements. Spinal cord injury is a well-known example of traumatic injury. With spinal cord
10 injury, the pathways that link higher motor centers in the brain with the spinal cord and that are used for control of voluntary movements can be functionally transected at the site of injury. As a result, the patient is paralyzed, and can no longer voluntarily activate muscles that are innervated by regions of the spinal cord below the level of the injury. Despite the injury to their long fibers,
15 however, many of the cells in these higher brain regions that control voluntary movement will survive and can still be activated voluntarily to generate electric signals for controlling voluntary movement. By recording the electrical activities produced from these cells with implantable neural sensors (e.g., a microwire electrode array, a microwire, a magnetic field detector, chemical sensor, or
20 other neural sensor), signals generated by the cells can be "exteriorized" and used for the control of external prostheses, such as an assist robot or an artificial limb, or functional electrical stimulation paralyzed muscles. Additionally, these generated signals can be used for control of computer operations such as the movement of a cursor on a computer display.

Another example of such loss occurs in cases of amyotrophic lateral sclerosis (Lou Gehrig's Disease), in which the motor neurons that control muscles, as well as some of the brain cells that control these motor neurons, degenerate. In advanced stages of this disease, the patient might have
5 completely intact senses and thought processes, but is "locked in," so that neither movements nor behavioral expressions of any kind can be made. Providing these patients with some way of communicating with the external world would greatly enhance their quality of life.

In sum, there is a need to develop a system for monitoring and
10 processing the electrical signals from neurons within the central nervous system, so that the brain's electrical activity can be "exteriorized" and used for the voluntary control of external prostheses or assist devices which are adapted to provide sensory feedback. In this way, damaged pathways can be circumvented and some control of the environment can be restored;
15 additionally, a patient can be provided the ability to interact with his or her environment. Because the electrical fields of small groups of neurons drop off rapidly with distance from the cells, a representative system can include surgically implanted electrodes or other neural sensors, which can be placed in close proximity to the individual or large numbers of brain cells that generate
20 command signals for voluntary movement.

Neural signals can be detected by measuring the electric field potential of an area or region of the brain or other organ. The field potential detected at any one point represents the sum of the potential created by a number of electric potential generators in the area surrounding the field potential

measuring device. By way of example, when an individual monitors a field potential (e.g., the amplitude of a field potential) at a point on the surface of the cerebral cortex, for example, what is detected is the overlapping summation of electric fields generated by active neurons in the depths of the cerebral cortex, which have spread through the tissues and up to the surface. These nerve cells can be characterized as point dipoles that are oriented perpendicular to the surface of the cerebral cortex. In other words, each cell or group of cells has a current source where positive charge moves outwardly across its membrane and a current sink where the same amount of positive charge moves inwardly at each instant. Thus, the flow of current across each cell or group of cells establishes an electric field potential that is equivalent to the electrostatic field potential of a pair of point charges, one positive at the location of the current source and one negative at the current sink. The amplitude of this field potential, i.e., the electric field strength, decreases inversely with distance in all directions from each point charge, and is relatively low at the surface of the cerebral cortex.

When many nerve cells are generating field potentials in a given region, these field potentials sum and overlap in the neural tissue, in the extracellular fluid, and at the brain surface. This summation is a linear function in this volume conductor, since the field strength of a given cell or group of cells varies inversely as a function of the distance from each current source or sink. Thus, if the electric potential of a given region of neurons is measured at a sufficient number of points and depths, it is possible to deduce the locations and amplitude of each dipole generator at any instant of time.

Integrated circuits, called neurochips, have been developed to acquire neural signals from a subject and condition the signals for processing. Some current neurochips include multi-channel sieve electrodes for detecting neural signals from regenerated axons. A sieve electrode is a planar structure with small throughbores extending therethrough. In order to implant a sieve electrode, an axon is severed, the ends placed through adjacent throughbores, and the nerve is allowed to heal. Signals in the regenerated axon are detected by the sieve electrode. Detected signals are then processed and transmitted by the neurochip for further processing.

Many current neural signal systems utilize radio frequency telemetry for transmitting information signals. A significant amount of the total power required for operating a neurochip is used to implement telemetry. High power consumption is undesirable for neurochips in order to achieve reduced neurochip and system size. Thus, neurochip telemetry and transmission methods are desired having lower power requirements for transmitting information signals. In general, double the power is required to transmit twice the amount of data. Thus, neurochips are desired that require as little data transmission as possible, thus using a lower amount of power to transmit. Further, neurochips are desired having a smaller size and improved circuitry for receiving, conditioning, and processing neural signals. Such improvements will reduce the amount of information that must be passed to through the telemetry links to other parts of the device thus conserving power and will distribute the processing burden to multiple devices operating serially and synchronously.

Disclosure of the Invention

According to one embodiment, a neural spike detection system is provided. The neural spike detection system can include a signal receiver operable to receive a plurality of neural signals comprising a neural spike. The system can also include a neural spike detector adapted to communicate with the signal receiver and detect the neural spike in the plurality of neural signals.

Further, the system can include a transmitter in communication with the neural spike detector and operable to transmit an information signal when a neural spike is detected.

According to a second embodiment, a neural signal detection system is provided. The neural signal detection system can include a signal receiver adapted to condition neural signals received from neural sensors. The system can also include a control module operable to select neural signals for transmission. Further, the system can include a transmitter operable to transmit the conditioned neural signals selected by the control module.

According to a third embodiment, a neural signal transmission system is provided. The system can include a signal receiver operable to condition a plurality of neural signals. The system can also include a wireless power receiver adapted to wirelessly receive power from a wireless power transmitter for powering the system. Further, the system can include a neural signal transmitter operable to transmit the conditioned neural signals.

According to a fourth embodiment, a method for transmitting a neural spike signal is provided. The method includes receiving a neural signal including neural spikes and detecting occurrences of neural spikes in the neural

signal. The method can also include transmitting an information signal indicating the occurrence of a neural spike when a neural spike is detected.

According to a fifth embodiment, a method for transmitting neural signals is provided. The method can include selecting neural signals received by
5 neural sensors for transmission. The method can also include conditioning the selected neural signals. Further, the method can include transmitting the neural signals.

According to a sixth embodiment, a method for transmitting neural signals is provided. The method can include conditioning a plurality of neural
10 signals. The method can also include receiving a wireless power signal from a wireless power transmitter for powering the system. Further, the method can include transmitting the neural signals.

According to a seventh embodiment, a neural processing system is provided. The system can include an implanted neurochip operable to transmit
15 a plurality of detected neural signals including neural spikes and noise. The system can also include a wearable relay device operable to receive the plurality of detected neural signals, filter the noise, and transmit a signal having the neural spike. Further, the method can include a remote processing system operable to receive the signal having the neural spike.

20 According to an eighth embodiment, a neural signal receiver for conditioning a plurality of neural signals is provided. The neural signal receiver can include a plurality of preamplifiers for conditioning the plurality of neural signals. The neural signal receiver can also include a differential amplifier module connected to the plurality of preamplifiers for selecting a reference

signal from among the plurality of neural signals and generating a plurality of difference signals. The plurality of difference signals can be a difference between the plurality of neural signals and the reference signal.

Some of the objects of the invention having been stated hereinabove,
5 other objects will become evident as the description proceeds when taken in connection with the accompanying drawings as best described hereinbelow.

Brief Description of the Drawings

Exemplary embodiments of the invention will now be explained with
10 reference to the accompanying drawings, of which:

Figure 1 is a schematic view of a neural signal system according to an embodiment of the present invention;

Figure 2 is a flow chart of a process for neural spike detection according to an embodiment of the present invention;

15 Figure 3 is a graphical view of simulation results of the input signal and output signal of a neural spike detector according to an embodiment of the present invention;

Figures 4A and 4B are a schematic views of a signal receiver having 16 channels according to an embodiment of the present invention;

20 Figure 5 is a schematic view of a signal conditioner module following a neural sensor according to an embodiment of the present invention;

Figures 6A and 6B are graphical views of actual and measured phase and gain responses of the embodiment of the present invention described with regard to Figure 5;

Figure 7 is a schematic view of a signal receiver having 16 channels according to an embodiment of the present invention;

Figure 8 is a schematic view of a preamplifier of signal conditioner module configured with an associated capacitor of a capacitor module
5 according to an embodiment of the present invention;

Figure 9 is schematic view of a signal receiver having 32 channels and communication with a remote computer system according to an embodiment of the present invention;

Figure 10 is a schematic view of a signal receiver having 16 channels
10 and communication with a remote computer system according to another embodiment of the present invention;

Figure 11 is a schematic view of a neural spike detector according to one embodiment of the present invention;

Figure 12 is a schematic view of a neural spike detector according to
15 one embodiment of the present invention;

Figure 13 is a schematic view of a signal transmitter according to an embodiment of the present invention;

Figure 14 is a schematic view of a signal receiver and a spike processor according to one embodiment of the present invention; and

20 Figure 15 is a schematic view of a signal receiver and a spike processor according to one embodiment of the present invention.

Detailed Description of the Invention

I. Definitions

Following long-standing patent law convention, the terms “a” and “an” mean “one or more” when used in this application, including the claims.

5 As used herein, the term “about,” when referring to a value or to an amount of mass, weight, time, volume, concentration or percentage is meant to encompass variations of $\pm 20\%$ or $\pm 10\%$, more preferably $\pm 5\%$, even more preferably $\pm 1\%$, and still more preferably $\pm 0.1\%$ from the specified amount, as such variations are appropriate to perform the disclosed method.

10 As used herein, the terms “actuator”, “external device” and “prosthetic limb” are used interchangeably and mean any kind of device adapted to perform a movement. Although an actuator preferably performs a movement in three dimensions, an actuator can also be limited to performing movements in two dimensions. Thus, an actuator can be a manipulandum confined to two-
15 dimensional motion. A representative actuator comprises a prosthetic limb, which can be fitted on, or integrated into, the body of a subject. An actuator can also be associated with machinery and/or circuitry that allow the actuator to respond to one or more forms of input with one or more movements. In one example, the range of motion of an actuator designated as a substitute for a
20 patient’s lost or paralyzed limb is limited to the range of motion of the limb for which the actuator is substituting.

As used herein, the term “electrode” means an electric conductor through which a voltage potential can be measured. An electrode can also be a collector and/or emitter of an electric current. In one embodiment, an

electrode is a solid and comprises a conducting metal. Representative conducting metals include noble metals, alloys and particularly stainless steel and tungsten. An electrode can also be a microwire, or the term "electrode" can describe a collection of microwires. In one embodiment, electrodes
5 comprise polytetrafluoroethylene (PTFE) (TEFLON®, a product produced by E.I. Du Pont de Nemours and Company) coated stainless steel or tungsten microwires.

As used herein, the terms "field potential data" and "field potentials" are used interchangeably and typically mean low frequency, voltage measurements
10 collected from one or more locations near or at one or more neurons in a subject's brain or nervous system.

As used herein, the term "integrated circuit" refers to a small-scale, electronic device densely packaged with more than one integrated, electrical component. The components are manufactured on the surface of
15 semiconductor material. There are various scales of integrated circuits that are classified based on the number of components per surface area of the semiconductor material, including small-scale integration (SSI), medium-scale integration (MSI), large-scale integration (LSI), very large-scale integration (VLSI), ultra large-scale integration (ULSI).

20 As used herein, the term "location source" means a position wherein a neural sensor can detect one or more neural signals.

As used herein, the term "neural signal" means a signal, which can take any form, originating in the nervous system of an organism. Neural signals

typically include neural spike signals that carry information in their arrival times at destination neurons.

As used herein, the term "neural sensor" means an implantable device for sensing neural signals. Examples of neural sensors include microwire
5 electrode arrays, optical sensors, microwires, magnetic field detectors, chemical sensors, and other suitable neural sensors which are known to those of skill in the art upon consideration of the present disclosure.

As used herein, the term "neurochip" means any integrated circuit, multi-integrated circuit, or multi-module system adapted for detecting neural signals
10 from the body of an organism. In one embodiment, a neurochip is adapted to be implanted in an organism near the nervous system for detecting neural signals.

As used herein, the terms "operator," "patient" and "subject" are used interchangeably and mean any individual monitoring or employing the present
15 invention, or an element thereof. Operators can be, for example, researchers gathering data from an individual, an individual who determines the parameters of operation of the present invention or the individual in or on which a high-density multichannel microelectrode array is disposed. Broadly, then, an "operator," "patient" or "subject" is one who is employing the present invention
20 for any purpose. As used herein, the terms "operator," "patient" and "subject" need not refer exclusively to human beings, but rather the terms encompass all organisms having neural tissue, such as monkeys, dogs, cats, rodents, etc.

II. General Considerations

Through the years there has been significant research in the area of detecting and observing various electric potentials generated within the human body for medical diagnosis, biofeedback control of mental and physical states, and control of external devices. It is known that different regions of the brain are used to control different parts of the body and to process different sensory inputs. It is also known that when a human performs a certain function, such as moving an extremity or listening to a particular sound, multiple regions of the brain generate electrical action potentials to accomplish that function. It is also known that direct electrical stimulation of a particular region of the brain can cause at least partial reproduction of the functions or sensory input normally associated with that region of the brain. Determining which portions of a patient's brain are responsible for certain motor activities or certain sensory functions has become known as brain "mapping." After a patient's brain has been mapped, the brain can be electrically stimulated to restore lost functions.

For example, it is possible to determine which portions of a patient's brain are responsible for processing signals associated with the movement of an extremity. Once a neurologist knows which portions of the patient's brain are responsible for processing these signals, it is possible to electrically stimulate selected portions of the patient's brain to cause the patient to "move" the extremity. Thus, a patient whose motor control has been partially or permanently damaged can regain motor control if an apparatus is employed to translate these neural signals into movement of an external device, such as an actuator. Similarly, if the areas of the patient's brain that are associated with

tactile and other sensory information are known, these areas of the patient's brain can be electrically stimulated to make the patient "experience" the sensory interaction between an object and an external device interacting with the object. Systems according to the present invention can be employed as a component of a system such as a closed loop brain-machine interface or intelligent brain pacemaker. These devices can greatly enhance the quality of life of individuals those individuals whose motor control has been impaired.

III. Configuration and Operation of the Neural Signal System

10 In accordance with the present invention, efficient methods and systems are provided for detecting, processing, and transmitting information contained in neural signals received from neural tissues, for example neural tissue of the brain or central nervous system. This stored information can be used to control an external device, such as an actuator, prosthetic device, or computer system, 15 or to treat a neurological condition. The methods and systems according to the present invention will be explained in the context of flow charts and diagrams. It is understood according to this invention that the flow charts and diagrams can be implemented in hardware, software, or a combination of hardware and software. Thus, the present invention can include computer program products 20 comprising computer-executable instructions embodied in computer-readable media for performing the steps illustrated in each of the flow charts or implementing the devices illustrated in each of the diagrams. Figure 1 is a schematic view of a neural signal system of the present invention, generally designated **100**, according to one embodiment of the present invention.

Broadly, neural signal system **100** includes the following main modules: a signal receiver **102**, a neural spike detector **104**, and a neural spike processor and transmitter **106**.

While in operation, signal receiver **102**, described in more detail below,
5 detects neural activity in the form of field potential data **108** generated from neural tissue (e.g., from large numbers of single neurons) in a subject, such as a human or a monkey. Signal receiver **102** converts field potential data **108** to an electrical-based representation of the neural signal that is suitable for processing by the hardware and/or software of system **100**. Signal receiver
10 **102** can include one or more neural sensors (e.g., metallic wire electrodes) implanted in a location in the subject for detecting field potential data **108** of the neural signals of interest. In one embodiment, signal receiver **102** can include multiple neural sensors that are positioned in different location sources in the subject for converting the field potential data detected from different neurons
15 into electrical-based neural signals.

Signal receiver **102** can also include electrical components for conditioning the electrical-based neural signals. Preferably, signal receiver **102** conditions the electrical-based neural signal with buffering and filtering to remove unwanted in-band and out-of-band signal noise, such as field potential
20 data detected from signals of other bioelectric generators in the subject. These sources include the heart (ECG), muscles (EMG), and signals from the effect of mechanical movement of the sensor in response to blood pressure, respiration, and physical motion. Further, signal receiver **102** can correct a DC offset problem associated with conditioning neural signals. Further, system **100** can

include a transmission link **110** for communicating amplified and conditioned, electrical-based, neural signals on one or more communication channels from signal receiver **102** to neural spike detector **104**.

Neural spike detector **104** can receive one or more conditioned signals
5 on one or more communication channels from signal receiver **102** for detecting the occurrence of a neural spike in any of the conditioned signals. Additionally, neural spike detector **104** can include spike sorting, the process of identifying which particular neuron detected on a conditioned signal produced a particular neural spike. Neural spike detector **104** can detect a neural spike and sort
10 detected spikes in a conditioned signal by performing a mathematical algorithm. In one embodiment, the mathematical algorithm can detect a neural spike on a conditioned signal by the following steps: (1) sampling the conditioned signal; (2) combining the samples; (3) comparing the combined samples to predetermined thresholds; and (4) determining whether a spike has
15 been detected based on the comparison in step (3). The comparison can be to one threshold value, for instance a voltage level, or could be multiple threshold values, for instance level and slope, or could be the level of a derived signal, for example, energy or absolute value, or the level of multiple derived signals. Similarly, the detector may differentiate between threshold crossings in the
20 positive direction and the negative direction. Upon detecting a neural spike, neural spike detector **104** can transmit an information signal to neural spike processor **106** to indicate the detection of a neural spike in one of the conditioned signals. Neural spike detector **104** can transmit information signals to neural spike processor **106** via transmission link **112**. When neural signals

are received from multiple channels or locations, the information signal can carry data identifying the neuron corresponding to a detected neural spike.

Neural spike processor **106** can receive the information signal from neural spike detector **104** and generate and transmit control signals based on the information signals. Control signals can be transmitted via a transmission link **114** to a device such as an actuator, prosthetic device, computer system, or other suitable device. Other devices include but are not limited to weapons or weapon systems, robots or robot systems, other commercial electronic devices that can be controlled remotely including TV, radio, mechanical bed systems stoves, ovens, and other cooking devices, other household devices that might be controlled by a remote device and used to improve the quality of life of a disabled person. Still other devices include scientific or commercial mechanical devices that work at a much larger or much smaller scale than is normal for a human, for instance optical tweezers for manipulating molecules and atoms, or earth moving equipment. Preferably, transmission link **114** comprises a wireless link such as ultra wide band (UWB) radio telemetry. Alternatively, the transmission link can comprise any other suitable wireless link such as by the BLUETOOTH™ standard developed by BLUETOOTH SIG, Inc.

Neural spike processor **106** can include a memory for storing information signals. Data can be stored in terms of the time the neural spike was generated and/or the channel or location source of the neural spike. The memory can also receive and store the signal data with information identifying the sensory or motor activity ongoing at the time of the neural spike. In one embodiment, the memory stores the signals in a digital format. Alternatively,

signal data can be represented by analog voltage records of the complete signal, a time multiplexed analog signal, analog records of only the neural spikes and a time indicator identifying the time of the neural spike, or any other suitable format known to those of skill in the art.

5 According to one embodiment, neural signal system **100** can also be operable to receive signals for configuring the components of system **100**, as described herein. Therefore, in this embodiment, links 110, 112, and 114 comprise bi-directional signals for passing configuration data.

10 In one embodiment, neural signal system **100** comprises a neurochip, an integrated device or a highly integrated package of semiconductor circuits adapted to receive signals from the neural tissue of a subject and process and condition neural signals. The neurochip can be implanted in the tissue of a subject or positioned outside the skin of the subject in a transcutaneous configuration. Alternatively, the neurochip can be positioned at a location on or
15 near the subject and can be adapted to interact with additional components, such as a signal receiver, via a conductive wire or wireless communication. The processing by the neurochip can be analog signal processing or digital signal processing. The neurochip can be implemented using a VLSI circuit. The use of integrated circuit technology allows the tailoring of device
20 parameters to optimally use the available power and space for the desired sensory or motor functions.

 In an alternative to the implementation of components **102**, **104**, and **106** on a single neurochip for implantation in a subject, signal receiver **102** and neural spike detector **104** can be implemented on a neurochip, and neural

spike processor **106** can be implemented separately as a non-implanted, integrated circuit. Further in the alternate, components **102**, **104**, and **106** of neural signal system **100** can be implemented together or separately on a single integrated circuit, any combination of one or more integrated circuits, or
5 any other suitable hardware and/or software combination known to those of skill in the art.

In one embodiment, signal receiver **102** is implemented in the form of an implant. Alternatively, one or more of components **104** and **106** can be implemented as an integrated circuit in the form of an implant, a
10 transcutaneous implant, a remote device, a portable remote device, a wearable device, or a wearable tethered device. An implant is a component residing completely within the subject. A transcutaneous implant is an implanted component having a mechanical transcutaneous link, such as a transcutaneous, conductive wire, or to a component located outside of the skin
15 of the subject. A remote device is a component remotely located from the subject. Typically, the remote device does not physically move with the subject. Alternatively, the remote device can move with the subject without being wearable by the subject. A wearable device is a component wearable by a subject, removed with no surgical procedure, and not physically connected to
20 a remote device. For example, a wearable device can be worn on the head, back, waist, or other convenient location on the subject (e.g., a helmet or backpack). A wearable, tethered device is a wearable device that is connected to a remote device via a wire link.

Components can be implemented as a combination of one or more implants, transcutaneous implants, remote devices, portable remote devices, wearable devices, or wearable tethered devices. In one exemplary embodiment, signal receiver **102** can be implemented as a transcutaneous
5 implant that communicates with a wearable, tethered neural spike detector **104** located on the head of a subject. In another exemplary embodiment, signal receiver **102** is implemented as an implant communicating via a wire link with a wearable neural spike detector **104**. Wearable neural spike detector **104** communicates with another wearable system including neural signal processor
10 **106** on the waist of a subject. Neural signal processor **106** can transmit via radio communication to a remote device for further interpretation.

As stated above, transmission links **110**, **112**, and **114** can be implemented as a wire link, a wireless link, or a combination of wire and wireless links. The following list includes exemplary embodiments of neural
15 signal system **100** having different wire and wireless link configurations according to the present invention:

- According to one embodiment, neural signal system **100** includes an implanted component, such as a neurochip, and a remote component, such as a processing system. The implanted component transmits
20 signals to the remote component via a wireless link.
- According to one embodiment, neural signal system **100** includes a wearable component, a transcutaneous implant, such as a neurochip, and a remote component, such as a computer processing system. The transcutaneous implant communicates with the wearable component via

a wire link. The wearable component communicates to the remote component via a wireless link.

- According to one embodiment, neural signal system **100** includes an implanted neurochip, a wearable component, and a remote component.

5 The implanted neurochip transmits signals to the wearable component via a wireless link. The wearable component transmits signals to the remote component via a wire link.

- According to one embodiment, neural signal system **100** includes an implanted component having a set of transcutaneous wires. System

10 **100** also includes a wearable component, such as a neurochip disposed in a helmet. The neurochip and implanted component communicate via a wireless link. Further, the wearable component transmits signals to a second wearable component via a wireless link. The second wearable component is tethered to a remote or portable remote component for
15 wire communication. Alternatively, the second wearable component can transmit control signals to an actuator.

- According to one embodiment, neural signal system **100** includes an implanted component, such as a neurochip, operable to transmit signals via a wireless link. The wireless link transmits signals optically to a
20 wearable component, such as a helmet. The wearable component is operable to relay signals to a remote component for further processing.

- According to one embodiment, neural signal system **100** includes a remote component operable to transmit data acquisition configuration parameters to an implanted component.

- According to one embodiment, neural signal system **100** includes an implanted neurochip for transmitting via a wireless link, such as optically, to a wearable component including a transmitter. The transmitter transmits via a wireless link, such as a radio link, to a remote component. The remote component transmits via a radio link to a wearable prosthetic device.
- According to one embodiment, neural signal system **100** includes an implanted component, such as a neurochip, for optically transmitting signals to a wearable component. The wearable component communicates with a wearable component, such as a prosthetic device, via a conductive wire.
- According to one embodiment, neural signal system **100** includes an implanted, transcutaneous component, such as a set of implanted transcutaneous wires, operable to transmit signals to a wearable component having a neurochip. The wearable component transmits control signals to a wearable component, such as a wearable prosthetic device.

III.A. Receiving Neural Signal

A method for receiving neural signals in neural signal system **100** according to the present invention can be performed by signal receiver **102**. The purpose of receiving neural signals is to convert a biological-based neural signal into an electrical-based, machine-readable signal and to condition the received signal. In one embodiment, neural signal detection according to the

present invention includes two main steps – (1) receiving the biological-based neural signal from a neuron of interest and (2) conditioning the neural signal. A transmission step optionally can be included if transmission link **110** is necessary for transmitting the conditioned signal from neural signal receiver

5 **102** to neural spike detector **104**.

The step of receiving the neural signal can include providing a neural sensor near a neuron to obtain a neural signal in the form of field potential data from a neuron or large number of single neurons. The neural sensor converts the field potential data to electrical-based current flow. The typical maximum

10 potential voltage associated with a propagating neuronal action potential is between approximately 100 and 700 microvolts (μV). The chemical reaction that performs this conversion requires a driving potential, referred to as polarization, that causes a significant direct current (DC) voltage of approximately 200 millivolts (mV) on the measured neural signals. Because

15 the offset voltage is a function of the ionic concentrations at the interface, small movements in the neural sensor associated with respiration, blood pressure, and locomotion can modulate the DC potential providing an additional source of electrical noise. The DC and low frequency noise components of the measured electrical-based signal are frequently orders of magnitude greater in

20 size than the neural signals of interest.

Isolating the neural signal is one of the primary challenges in building an integrated circuit for this purpose. The neural sensor in the subject can be situated among thousands of neurons and thus measures neural spikes from many neurons simultaneously. Only a few neurons produce signals with

amplitude large enough to be accurately interpreted. The remaining neurons produce signals that are interpreted as noise. The neural information carried by the neural spike is contained in its timing relative to other neural spikes from the same neuron. Thus, it is important to be able to differentiate spikes
5 originating from a specific neuron or neurons. Neural spike sorting schemes for identifying the neuronal origin of a given neural spike can be implemented in hardware and/or software. Typically, neural spike sorting schemes require the entire temporal waveform of the approximately 1-2 millisecond (ms) neural spike to be acquired and analyzed by the system.

10 Conditioning the electrical-based signal can include the steps of filtering and differential recording the detected signal. As stated above, the electrical signal of interest is a neural spike approximately 2 milliseconds (ms) in duration. The frequency content of the neural spike can be between approximately 500 and 8000 Hertz. This relatively small voltage spike is
15 corrupted by several sources of noise both internal and external. Some significant sources of corruption is from other bioelectric generators within a subject, such as cardiac, neuromuscular, and other cortical neurons located in proximity to the electrodes. Filtering and differential recording eliminates sources of out-of band noise or noise in frequencies outside the frequencies of
20 interest. Differential recording eliminates in-band noise common to both the signal and its reference voltage.

III.B. Neural Spike Detection

A method for neural spike detection in neural spike detector **104** according to the present invention includes receiving conditioned neural signals and detecting neural spikes in the conditioned signals. Neural spike detector **104** can also include sorting spikes according to the particular neuron producing a detected spike. As discussed above, neural spike detector **104** can detect neural spikes among noise signals by threshold and slope detection. In one embodiment, neural spike detection includes four main steps – (1) sampling the conditioned signal; (2) combining the samples; (3) comparing the combined samples to predetermined thresholds; and (4) determining whether a spike has been detected based on the comparison in step (3). One such process can digitally filter the signal to produce a measure of slope and detect the location of a slope above or below a particular value that corresponds in time with a voltage level exceeding a particular value.

Referring to Figure 2, a flow chart is provided which illustrates a process for neural spike detection according to an embodiment of the present invention.

As stated above, such neural spike detection can be performed by neural spike detector **104**. The process begins at the step indicated by reference numeral **202**. At step **204**, neural signals are received by neural spike detector **104**. The received neural signals can be electrical-based analog signals. Next, at step **206**, the neural signals can be sampled N times. In one embodiment, the neural signal is sampled 8 times at 40 kHz. The N samples can be combined at step **208**. In one embodiment, the samples are combined using gain. Alternatively, the samples can be combined by summing, power

operations or other suitable methods known to those of skill in the art. Next, at step **210**, it is determined whether the samples are greater than a predetermined threshold. The comparisons can be made to determine the presence of a particular feature of a spike, for example, the point at which the slope or the amplitude or the energy exceeds a threshold. This point can then be taken as the time the spike occurred or some temporal displacement from that point is taken as the time the spike occurred (i.e., 1 millisecond earlier). If the samples are greater than the predetermined threshold, a spike is detected (step **212**). If the samples are not greater than the predetermined threshold, a spike is not detected (step **214**). Next, the process stops (step **216**). The process can then repeat for another sample group at step **202**.

Referring to Figure 3, a graphical view of computer simulation results are provided of the input signal and output signal of a neural spike detector **104** according to one embodiment of the present invention. The vertical axis indicates the amplitude of a neural signal. The horizontal axis indicates time. In this embodiment, neural spike detector **104** can determine whether a neural spike is starting to occur based on a number of samples of a neural signal. First, neural spike detector **104** samples a neural signal, obtaining 8 sequential values. Next, neural spike detector **104** determines a neural spike is starting if the sum of the last two samples is less than 0.23 times the sum of the first six samples. Reference numeral **302** indicates the neural signal input to neural spike detector **104**. Reference numerals **304** indicates instances of spike detection. In the simulation of this embodiment, 99.97% of all neural spikes in the simulation were detected. Neural spike detector **104** of this embodiment

indicated neural spikes falsely about half of the time. These computer simulations were conducted using the MATLAB® mathematics computer program, produced by MathWorks, Inc. of Natick, Massachusetts.

Neural spike detection and sorting can be performed with either analog
5 or digital circuitry. An integrated monolithic analog circuit can detect spikes and pass the reduced data on to a sorter before or after a telemetry link. Alternatively, a monolithic analog circuit can process the detected spikes and perform spike sorting. Further, in the alternative, spike detection and sorting can be performed by digital circuitry by digitizing the neural signals and
10 processing the resultant digital data stream using a digital signal processor or a custom digital circuit.

Upon detecting a neural spike, neural spike detector **104** can indicate the detection of a neural spike on one or more of the communication channels to neural spike processor **106**. If necessary, neural spike detector **104** can
15 transmit an information signal to neural spike processor via transmission link **112**. When neural signals are received from multiple channels or locations, the information signal can carry data identifying the channel or location source corresponding to the detected neural spike. Further, if spike sorting is performed before transmission on transmission link **112**, the information signal
20 can carry information indicating the particular neuron originating the neural spike.

III.C. Neural Spike Processor, Transmitter, and Receiver

A method for neural spike processing and transmission according to the present invention can be performed by neural spike processor **106**. The purpose of neural spike processing and transmission is to generate and
5 transmit information or control signals to a suitable device such as an actuator, prosthetic device, or computer system. In one embodiment, the information or control signals are transmitted via a wireless link.

Signal processing of various types can occur. This processing is divided into two stages: 1) processing within one signal 2) processing across many or
10 all signals. In the first stage, the signals can be used with sorting or without sorting. A sorting process generally increases the number of signals by separating portions of one signal into one or more additional signals of spike data. In the second stage, an estimate of firing rate can be made based on a summing of spike counts over an interval (binning) for example at 100
15 milliseconds. Other methods of estimating rate are familiar to those skilled in the art. Similarly, a process can use the temporal order of the signals to derive other parameters for use in control processing. Such time based interpretation schemes are also known to those skilled in the art. After extraction of the information from a single channel of spikes, the information among signals can
20 be combined. The signals can be combined with linear weighting schemes or using techniques such as neural networks.

As stated above, component **106** can communicate over transmission link **114** via a wireless link. A wireless link is desirable because a subject can move freely within the transmission range of the wireless link. In one

embodiment, the wireless link is implemented with ultra wide band (UWB) radio. UWB radio is particularly suited to digital data transmission at high bandwidth (data rates) over short distances. The advantages of UWB radio over other communication systems include low power per bit of information and
5 simplicity of transmission electronics.

Information signals can be transmitted via UWB radio in real-time for maintaining the neural spike sequence. For example, if 50 channels are monitored, each time a spike is detected on a channel, a number identifying the channel or origin of the neural spike is transmitted via UWB radio.
10 Transmission can include a short delay due to processing. An indication of the channel or origin of the neural spike is transmitted via a short identification transmission. Short identification transmission resolves situations in which two or more neurons fire at the same time. For example, if identification requires 200 nanoseconds and a neural spike lasts 2 milliseconds (ms), the system has
15 the capacity for transmitting the identification of 10,000 neural spikes. In a neural signal system having time designated to the nearest millisecond, the system can transmit identification for 5,000 neural spikes per millisecond.

The method of encoding spikes for radio transmission requires minimal power because the temporal firing patterns are simply reproduced by the radio
20 transmission and the time of firing is inherent within the transmission structure. Thus, information regarding the time of a neural spike is not required, only a neural channel or origin identifier. This type of coding reduces the required bandwidth and reduces power consumption by more than 50%. Power consumption is further reduced because information is transmitted only when a

neural spike is detected and the time required to transmit that information is very short compared to the intervals between neural spikes on a given channel. For example, if it takes 200 nanoseconds to transmit an information signal indicating a neural spike, a typical 50 Hertz firing rate for a given neuron would
5 result in a duty cycle of only approximately 0.001 per channel, assuming one neuron per channel. All analog signal processing and spike sorting is performed prior to transmission on UWB transmitter.

As stated above, neural spike processor and transmitter **106** can include a memory for storing information signals. Data can be stored in terms of the
10 time the neural spike was generated and/or the channel or location source of the neural spike. The memory can also receive and store the signal data with information identifying the sensory or motor activity ongoing at the time of the neural spike. Preferably, the memory stores the signals in a digital format. Alternatively, signal data can be represented by analog voltage records of the
15 complete signal, a time multiplexed analog signal, analog records of only the neural spikes and a time indicator identifying the time of the neural spike, or any other suitable format known to those of skill in the art.

IV. Embodiments of a Neural Signal System

20 IV.A. Signal Receiver

Referring to Figures 4A and 4B, schematic views of a signal receiver, generally designated **400**, having 16 channels according to an embodiment of the present invention is illustrated. Signal receiver **400** can include a signal conditioner module **402** and an analog-to-digital (A/D) converter module **404**.

Signal receiver **400** is operable to receive neural signals from a first group of neural sensors **406** (NS1 – NS8) and a second group of neural sensors **408** (NS9 – NS16), condition the neural signals, and provide a digital representation of the neural signals to a neural spike detector or other hardware and/or software combination for further processing such as neural spike detection, sorting, display, or control of a prosthetic device. Signal conditioner module **402** can include input channels for receiving neural signals and connecting first and second group of neural sensors **406** and **408** to a first group of preamplifiers **410** (P1 – P8) and a second group of pre-amplifiers **412** (P9 – P16), respectively, for providing high gain input amplification and band pass filtering for the neural signals of first and second group of neural sensors **406** and **408**, respectively. Preamplifiers **410** and **412** can be followed by a differential amplifier module **414** having a first and second multiplexer **416** and **418** and switches (not shown) for selecting two neural sensors from first and second group of neural sensors **406** and **408**, respectively, for providing reference signals to eliminate common mode noise, described in more detail below. Such signal could come from a standard neural sensor or from a sensor designed specifically to act as a reference electrode. Further, differential amplifier module **414** can include a switch **420** connected between the outputs of multiplexer **416** and **418** for selecting any of neural sensors **406** and **408** to provide a reference signal. Differential amplifier module **414** can also include switches **422** and **424** connected between the output of multiplexers **416** and **418**, respectively, and ground for selectively grounding the reference signal from either multiplexer **416** or multiplexer **418**. Differential amplifier module

414 can also include differential amplifiers **426** for providing a difference signal that between a neural signal and one of the reference signals for eliminating common mode noise signals. Differential amplifiers **426** can also provide additional gain to the neural signals. In one embodiment, multiplexers **416** and **418** comprise ANALOG DEVICES™ AD708 multiplexer produced by Analog Devices, Inc. of Norwood, Maine. In one embodiment, switch **420**, **422**, and **424** is a MAXIM™ 4626 switch produced by Maxim Integrated Products, Inc. of Sunnyvale, California.

Signal conditioner module **402** can include filters **428**, described in more detail below, connected to differential amplifiers **426** for providing high and low pass filtering to the difference signals. In one embodiment, filters **428** can be adjusted to filter different frequency ranges. Filters **428** can be followed by variable gain amplifier **430** for selectively adjusting the gain of the signals. In the case of analog-to-digital conversion in a later stage, the gain adjustment can reduce the number of bits of resolution required in the analog-to-digital converter.

Signal receiver **400** can also comprise a control module **436** for controlling various components of signal conditioner module **402** and A/D converter module **404**. Control module **404** can be an autonomous system or an operator-assisted system. In the case of an operator-assisted system, the operator visualizes each signal in turn and decides if the signal is an appropriate neural signal or is noise and turns the channel on or off depending on this decision. The operator can select the appropriate reference electrode by looking for one channel that shows little or no discrete activity and would

thus be suitable as a reference. Similarly, the operator can optimize the gain on a particular channel based on the size of the neural signals. In the case of an autonomous or semi-autonomous system, a computer program can search for similar criterion as described above for selecting the proper control settings.

5 For instance, the rate of neural spiking as determined by the detector can be an indication if the signal had an appropriate neural signal or was just measuring noise. For automatic gain selection, an autonomous device can monitor the signal levels and adjust the gain so that the fall within a specific voltage range suitable for processing by the A/D converter or other following
10 stage. Control module **404** can receive signals from a processor (not shown) for controlling the components of signal conditioner module **402** and A/D converter module **404**. Control module **404** can be implanted into a subject or worn by the subject.

Control module **402** can comprise a channel on/off module **438**
15 connected to preamplifiers **410** and **412** for selectively powering preamplifiers **410** and **412**. Module **438** can turn on/off each preamplifier of preamplifiers **410** and **412** for conserving power when it is determined that a corresponding one of neural sensors **406** and **408** is receiving a suitable neural signal from the subject. Selector **438** is shown in Figures 4A and 4B connected to only one of
20 preamplifiers **410** and **412** for clarifying the illustration.

Control module **436** can comprise a reference channel selector **440** connected to multiplexers **416** and **418** and switches **420**, **422**, and **424** for choosing a neural signal or ground as a reference signal. Selector **440** can select one of the inputs into multiplexers **416** and **418** as an output to function

as a reference signal. Additionally, selector **440** can control switches **422** and **424** to close to provide ground as a reference signal or open to provide for the output of multiplexers **416** and **418** as the reference signal. Switch **420** can be controlled to select any of neural sensors **406** and **408** as a reference signal as
5 described above.

Control module **436** can comprise a filter frequency selector **442** connected to filters **428** for selectively adjusting the frequency ranges filtered by filters **428**. The frequency ranges can be selected by adjusting either the capacitive or resistive components of filters **428**. The capacitive elements
10 could be adjusted using electronic switches to add or delete discrete values of capacitance. The resistive components of the filter circuit can be adjusted using a digital potentiometer or another suitable adjustable resistive component known to those of skill in the art. If the filters **428** are composed of integrated switched capacitor elements, the frequency can be adjusted by modifying the
15 clock frequency used to drive the switched capacitor elements. Selector **442** is shown in Figures 4A and 4B connected to only one of filters **428** for clarifying the illustration.

Control module **436** can comprise an amplification selector **444** connected to amplifiers **430** for selecting a magnitude of amplification for each
20 of amplifiers **430**. The amplification can be adjusted by modifying the resistive elements of the amplifier circuit or in the case of a switched capacitor amplifier by adjusting the clock frequencies. Selector **444** is shown in Figures 4A and 4B connected to only one of amplifiers **410** and **412** for clarifying the illustration.

Control module **436** can comprise a channel selector **446** connected to multiplexer **432** for selecting the channels output to A/D converter **434**. Multiplexer **432** can control multiplexer **432** to output only the channels carrying suitable neural signals.

5 Referring to Figure 5, a schematic view of components in signal conditioner module **402** following a neural sensor **406** according to an embodiment of the present invention is illustrated. Signal receiver **400** can be include a gain to vary between 3,200 and 50,000. Signal receiver **400** can also have bandpass filter corner frequencies of about 350 hertz and 8,000 hertz.

10 Small amplitude noise signals detected by neural sensors **406** and **408** necessitate low noise and high gain preamplifiers **410** and **412**. Noise can be limited by combining a low noise operational amplifier and a low-pass filter. Preamplifier **410** can include a capacitor **C1 500** and a resistor **R1 502** connected to neural sensor **406** and ground **504** for providing a unity gain,

15 high-pass filter. The output of the unity gain, high-pass filter is followed by a non-inverting, low-pass filter having gain. The high-pass filter can include an operational amplifier **506**, a capacitor **C2 508**, and resistors **R2 510** and **R3 512**. The output of the low-pass filter can be connected to the non-inverting input of operational amplifier **506** to prevent loading of the high-pass filter.

20 Resistor **R2 510** and capacitor **C2 508** are connected in a feedback configuration from the output of to the inverting input of operational amplifier **506**. Resistor **R3 512** is connected between ground **504** and the inverting input of operational amplifier **506**. In one embodiment, operational amplifier **506** is a MAXIM™ 4253 operational amplifier produced by Maxim Integrated Products,

Inc. of Sunnyvale, California for featuring low noise and shutdown to reduce power consumption when not operational. In this embodiment, the gain of preamplifier **410** is 100.

As stated above, differential amplifier module **414** can follow
5 preamplifiers **410** and **412**. Differential amplifier module **414** follows preamplifiers **410** and **412** for applying more gain in an early stage and improving signal quality by allowing for the rejection of common-mode noise signals. Module **414** can include a differential amplifier **514** (indicated with broken line). In this embodiment, differential amplifier **514** is a MAXIM 4199
10 differential amplifier produced by Maxim Integrated Products, Inc. of Sunnyvale, California for providing a differential gain with a common mode rejection ratio (CMRR) of about 110 decibels. Differential amplifier **514** can include an operational amplifier **516**, resistors **Rd1 518** and **Rd2 520** of the inverting input and resistors **Rd1 522** and **Rd2 524** of the noninverting input. The output of
15 preamplifier **410** is connected to resistor **Rd1 518** of differential amplifier **514**. The reference signal selected by multiplexers **416** and **418** (shown in Figures 4A and 4B) is connected to the resistor **Rd1 522** of differential amplifier **514**. Module **414** can also include an operational amplifier **526**, a capacitor **C3 528**, and a resistor **R4 530** configured as an integrator in the feedback of differential
20 amplifier **514**. The noninverting input of operational amplifier **526** is connected to ground **504**. Resistor **R4 530** is connected between the inverting input of operational amplifier **526** and resistor **Rd2 520** of differential amplifier **514**. Capacitor **C3 528** is connected between the inverting input of operational amplifier **526** and resistor **Rd2 524** of differential amplifier **514**. Placement of

the integrator in the feedback path of differential amplifier **514** improves the common mode range by adding a high-pass pole to attenuate low frequency offset voltages. In one embodiment, operational amplifier **526** is Burr Brown OPA2244 operational amplifier produced by Texas Instruments Incorporated.

5 In one embodiment, filter **428** is configured with Bessel filters. Bessel filters can preserve the activation signals if any waveform-based spike sorter is used. Filter **428** can include one or more cascaded filters: (1) a high-pass filter **530**, (2) a first low-pass filter **532**, and (3) a second low-pass filter **534**. Low pass filter **530** can include an operational amplifier **536**, capacitors **C4 538** and
10 **C5 540**, and resistors **R5 542**, **R6 544**, **R7 546**, and **R8 548**. Capacitors **C4 538** and **C5 540** are connected in series between the noninverting input of operational amplifier **536** and the output of operational amplifier **516** of differential amplifier module **414**. Resistor **R6 544** is connected from the output of operational amplifier **536** to the node connecting capacitor **C4 538** and **C5**
15 **540**. Resistor **R5 542** is connected between ground **504** and the noninverting input of operational amplifier **536**. Resistor **R7 546** is connected between ground **504** and the inverting input of operational amplifier **536**. Resistor **R8 548** is connected between the output and the inverting input of operational amplifier **536**.

20 First low-pass filter **532** can include an operational amplifier **550**, capacitors **C5 552** and **C6 554**, and resistors **R9 556**, **R10 558**, **R11 560**, and **R12 562**. Resistors **R9 556** and **R10 558** are connected in series between the noninverting input of operational amplifier **550** and the output of operational

amplifier **536** of high-pass filter **530**. Capacitor **C5 552** is connected from the output of operational amplifier **550** to the node connecting resistors **R9 556** and **R10 558**. Capacitor **C6 554** is connected between ground **504** and the noninverting input of operational amplifier **550**. Resistor **R11 546** is connected
5 between ground **504** and the inverting input of operational amplifier **550**. Resistor **R12 562** is connected between the output and the inverting input of operational amplifier **550**.

Second low-pass filter **534** can include an operational amplifier **564**, capacitors **C7 566** and **C8 568**, and resistors **R13 570**, **R14 572**, **R15 574**, and
10 **R16 576**. Resistors **R13 570** and **R14 572** are connected in series between the noninverting input of operational amplifier **564** and the output of operational amplifier **550** of first low-pass filter **532**. Capacitor **C7 566** is connected from the output of operational amplifier **564** to the node connecting resistors **R13 570** and **R14 572**. Capacitor **C8 568** is connected between ground **504** and
15 the noninverting input of operational amplifier **564**. Resistor **R15 574** is connected between ground **504** and the inverting input of operational amplifier **564**. Resistor **R16 576** is connected between the output and the inverting input of operational amplifier **564**.

Since Bessel filters can tend minimize phase distortion at the expense of
20 sharp filter roll-offs, higher order filters can be used. In the embodiment described with regard to Figure 5, there are four high-pass and five low-pass poles. Referring to Figures 6A and 6B, graphical views of actual and measured phase and gain responses of the embodiment described with regard to Figure 5. Referring specifically to Figure 6A, the gain response in the vertical axis is

shown verses the frequency spectrum in the horizontal axis. Referring specifically to Figure 6B, the phase response in the left vertical axis and the group delay in the right vertical axis is shown verses the frequency spectrum in the horizontal axis. The maximum group delay at any frequency is about 2
5 milliseconds.

In one embodiment, high-pass filter **532**, first low-pass filter **534**, and second low-pass filter **536** can comprise three Sallen-Key filters for forming Bessel filters. The operational amplifiers in the three Sallen-Key filters can be a BURR-BROWN™ OPA4244 quad package produced by Texas Instruments
10 Incorporated.

Variable gain amplifier **430** can be positioned after filters **428** for driving the capacitive input load of multiplexer **432**. Variable gain amplifier **430** can include an operational amplifier **578**, resistor **R17 580**, and a variable resistor **582**. Variable gain amplifier **430** can increase the dynamic range of signal
15 conditioner module **402**. The output of operational amplifier **465** of second low-pass filter **534** can be connected to the noninverting input of operational amplifier **578**. Resistor **R17 580** can be connected between ground **504** and the inverting input of operational amplifier **578**. Variable resistor **VR 582** can be connected between the output and inverting input of operational amplifier **578**
20 for controlling the gain of the input signal to variable gain amplifier **430**. In one embodiment, variable resistor **VR 582** is a 200 kilohm potentiometer. Alternatively, variable resistor **VR 582** can be a MAXIM™ 5160 potentiometer produced by Maxim Integrated Products, Inc. of Sunnyvale, California. The MAXIM™ 5160 is a 200 kilohm, 32-tap digital potentiometer evenly spaced

between one and 16.5. In one embodiment, operational amplifier **578** is available on the BURR-BROWN™ OPA4244 quad package for use in variable gain amplifier **430**.

Referring again to Figures 4A and 4B, A/D converter module **404** can
5 include a time-division multiplexer **432** and an A/D converter **434** for providing time-division multiplexing and sampling, respectively. Time-division multiplexer **432** can include 16 inputs for connection to the outputs of variable gain amplifiers **430**. In one embodiment, time-division multiplexer **432** is an ANALOG DEVICES™ AD706 multiplexer produced by Analog Devices, Inc. of
10 Norwood, Maine. In one embodiment, A/D converter **434** includes 12-bit, low power, successive approximation analog-to-digital conversion with a maximum throughput rate of 1 mega samples per second for allowing each of the 16 channels to be sampled at up to 62.5 kilo samples per second. In one embodiment, A/D converter **434** is an ANALOG DEVICES™ AD7495 analog-
15 to-digital converter produced by Analog Devices, Inc. of Norwood, Maine. In one embodiment, timing signals for time-division multiplexer **432** and A/D converter **434** are produced on a separate integrated circuit board. The output of A/D converter **434** can be transmitted to a processor, such as a processor of computer system **902** shown in Figure 9.

20 In one embodiment, signal conditioner module **402** and A/D converter module **404** are manufactured on the same integrated circuit board. Alternatively, signal conditioner module **402** and A/D converter module **404** can be manufactured on different boards. Power to signal conditioner module **402** and A/D converter module **404** can be supplied by voltage regulators (not

shown). In one embodiment, the voltage regulators are two high precision, low dropout voltage regulators. The two voltage regulators can be ANALOG DEVICES™ REF191 and REF198 produced by Analog Devices, Inc. of Norwood, Maine, which generate 2.048 volts and 4.096 volts, respectively. The

5 2.048 voltage rails of ANALOG DEVICES™ REF191 can be used as a virtual ground for effectively creating a ± 2.048 volt power supply without requiring an inverting voltage regulator or negative power supply. Since the ANALOG DEVICES™ REF191, by acting as the virtual ground, is incapable of sinking current, its output can be buffered with a unity gain source follower, for example

10 an ANALOG DEVICES™ OP262 produced by Analog Devices, Inc. The virtual ground can be used to ground the subject, provided that the power source is isolated from true ground.

In one embodiment, signal receiver **400** comprises a six-layer printed circuit board (PCB) for holding the components of signal receiver **400** and a

15 power supply. The PCB can include two 30-pin digital input/output (I/O) connectors and a 20 pin analog input zero input force (ZIF) connector. The analog channels can be laid out in eight parallel rows on each side of a PCB. A high board density can be achieved by using the smallest available hand-solderable parts, such as size 0402 for the passive components, 6.25 mil trace

20 widths, and 24 mil vias with 10 mil drill holes. The analog input connector can include the 16 input lines and two power lines and two ground lines for powering an active circuit. Preamplifiers **410** and **412** and differential amplifier module **414** can be protected from electromagnetic radiation by a grounded metal shield positioned over both sides of the PCB.

In one embodiment, a PCB having signal receiver **400** and a power supply includes 50 digital control signals listed in Table I below:

Control Signal	Quantity
ADC Clock	1
ADC Chip Select	1
Variable Gain Up/Down	1
Ground Select Switches	3
Multiplexer 416 Sel/Enb	4
Multiplexer 418 Sel/Enb	4
Multiplexer 432 Sel/Enb	4
Channel Enable	16
Variable Gain Clocks	16

- 5 The ADC clock and chip select control signals can control A/D converter **434**. Variable gain up/down control signal can control the resistance of variable resistor VR 582 for adjusting the gain of variable gain amplifier **430**. Ground select switches control signals can control switches **420**, **422**, and **424**. Multiplexer **416** sel/enb, multiplexer **418** sel/enb, and multiplexer **432** sel/enb
- 10 control signals can control multiplexers **416**, **418**, and **420**, respectively. Channel enable and variable gain clocks control signals can control amplifier enable signals and the variable gain amplifiers in each channel.

Digital input/output can be handled through two parallel 30-pin connectors. The connectors can have 50 control signals, four reference power

15 supply lines, four ground lines, and the output of A/D converter **434**. In one embodiment, the connectors can be connected to a remote computer or processor for controlling and receiving signals from signal receiver **400**. A/D converter **434** can be connected to a transmitter **448**, as described herein, for

transmitting the output of converter **434** to the remote computer or processor. Similarly, a remote transmitter can transmit control signals to a receiver **450** that are passed to control module **436** and used to set the state of selectors **438, 440, 442, 444, and 446**.

5 Referring to Figure 7, a schematic view of a signal receiver, generally designated **700**, having 16 channels according to another embodiment of the present invention is illustrated. Signal receiver **700** can include a signal conditioner module **702**, a capacitor module **704**, and an A/D converter module **706**. Signal receiver **700** is operable to receive neural signals from a first group
10 of neural sensors **708** (NS1 – NS8) and a second group of neural sensors **710** (NS9 – NS16), condition the neural signals, and provide a digital representation of the neural signals to a neural spike detector or other hardware and/or software combination for further processing such as neural spike detection, sorting, display, or control of a prosthetic device. Signal conditioner module
15 **702** can include input channels for receiving neural signals and connecting first and second group of neural sensors **708** and **710** to a pre-amplification stage **712** (indicated with broken line) having a first group of preamplifiers **714** (PREAMP1 – PREAMP8) and a second group of preamplifiers **716** (PREAMP9 – PREAMP16) for providing high gain input amplification and band pass filtering
20 for the neural signals of first and second group of neural sensors **408** and **410**, respectively. Pre-amplification stage **712** can be followed by a differential amplifier stage **718** (indicated with broken line) having a first and second multiplexer **720** and **722** for selecting two neural sensors from first and second group of neural signals **708** and **710**, respectively, for providing reference

signals to eliminate common mode noise, described in more detail below. Differential amplifier stage **718** can also include operational amplifiers **724** for providing a difference signal between a neural signal and one of the reference signals for eliminating common mode noise signals. Capacitor module **704** can
5 include 16 capacitors **726** connected between a ground **728** and preamplifiers **714** and **716** for reducing the effect of DC offset.

Referring to Figure 8, a schematic view of a preamplifier **714** of signal conditioner module **702** configured with an associated capacitor **726** of capacitor module **704** (shown in Figure 7) according to an embodiment of the
10 present invention is illustrated. Components of first group of preamplifiers **714** and second group of preamplifiers **716** are the same in this embodiment. As stated above, each preamplifier **714** is connected to one of capacitors **726** of capacitor module **704**. Preamplifier **714** can include an operational amplifier **800** and resistors **R1 802** and **R2 804**. Resistor **R1 802** is connected between
15 the inverting input of operational amplifier **800** and ground **728**. Resistor **R2 804** is connected between the output and inverting input of operational amplifier **800**. In one embodiment, the configuration of operational amplifier **800** and resistors **R1 802** and **R2 804** provide a gain of 50. According to one embodiment, resistor **R1 802** is 7.3 kilo ohms and **R2 804** is 357 kilo ohms.
20 Capacitor **726** is connected in the feedback of operational amplifier **800** between resistor **R1 802** and ground **728** to provide a high-pass portion for forcing the gain of pre-amplification stage **712** to unity for DC signals. According to one embodiment, capacitor **726** is 100 nanofarads. The low-pass

filter is provided by the gain bandwidth product of operational amplifier **800**. Capacitor **726** provides correction of the DC offset problem associated with conditioning neural signals. Input from the neural signal can be connected to the noninverting input of operational amplifier **800**. In this embodiment, pre-amplifier **714** provides a gain of 50 and band pass filtering between 218 hertz and 10,000 hertz to the received neural signal.

Preferably, signal conditioner module **702**, capacitor module **704**, and A/D converter module **706** are manufactured on separate integrated circuits. This design for pre-amplifier **714** has the advantage of high input impedance and large pass band gain and only one off chip component, capacitor **726**, for signal conditioning and only one additional input/output (I/O) pin required to connect capacitor **726** to a preamplifier of first group of preamplifiers **714** or second group of preamplifiers **716**. The circuit operation offers the advantage of reducing the effect of DC offset on a very broad range of neural sensors. This configuration can present a very high input impedance to the sensor thus allowing sensors with a high or low output impedance to be monitored. Manufacturing the modules on separate chips can allow commercial devices to be used for the capacitor module and the A/D converter modules. These parts can be made from different IC technologies and manufacturing them on different chips can allow the optimum technology to be used for each module.

Referring again to Figure 7, as stated above, differential amplifier stage **718** can also include first and second multiplexers **720** and **722** for providing selection of a neural signal from a first group of preamplifiers **714** (PREAMP1 – PREAMP8) and a second group of preamplifiers **716** (PREAMP9 –

PREAMP16) as reference signals to eliminate common mode noise. As stated above, signal receiver **700** can include inputs for 16 neural sensors **708**, NS1 to NS16. In this embodiment, the neural sensors are divided into a first group (NS1 – NS8) **708** and a second group (NS9 – NS16) **710** for positioning in generally different locations of the neural tissue of a subject. First group (NS1 – NS8) **708** and second group (NS9 – NS16) **710** can be connected to first multiplexer **720** and second multiplexer **722**, respectively, for selectively setting a reference signal for the neural signals of the respective group. Taking a reference signal from the portion of neural tissue close to the other sensors allows for the optimal cancellation of signals common to both sensors. These common signals are typically noise from both intrinsic and extrinsic sources. First multiplexer **720** and second multiplexer **722** include 8 inputs for connection to the outputs of first and second preamplifiers **714** and **716**, respectively, and can be set to one of the neural signals of the respective group as the reference signal for output. Preferably, the reference signal is set to a suitable neural signal. A reference signal can contain all of the noise signals common to the neural signal but does not contain the neural signal itself. Thus, when the reference signal is subtracted from signal recorded from the neural electrode, all of the noise will be removed. A good reference signal can be purposely created by implanting a sensor specifically to be used as a reference sensor, or a reference signal can be selected from among the available neural signals by examining the signal properties of each one using a distant reference signal and determining which of the neural signals is composed of only noise signals and contains little or no components from discrete neural

generators. In the alternative, signal receiver **700** can include circuitry for automatically selecting and setting suitable reference signals. The output of first multiplexer **720** and second multiplexer **722** can be connected to the inverting input of each of operational amplifiers **724** in its associated group of

5 operational amplifiers. The outputs of preamplifiers **714** and **716** can be connected to one of the noninverting inputs of its associated operational amplifiers **724**. Therefore, the resulting output of operational amplifiers **724** is the difference signal of the inputs. The difference signal is the difference between the input neural signal from one of preamplifiers **714** and **716** and the

10 reference signal of one of multiplexers **720** and **722**. Operational amplifiers **724** can also provide a gain to the difference signal. In one embodiment, operational amplifiers **724** provide a gain of 10.

A/D converter module **706** can include a third and fourth multiplexer **730** and **732** for multiplexing the difference signals at the outputs of operational

15 amplifiers **724** into two signals for analog-to-digital conversion by A/D converters **734**. The outputs of A/D converters **734** is the digital representation of the difference signal. Alternatively, module **706** can include a number of multiplexers up to one half of the total number of channels and a number of A/D converters **734** up to the total number of channels. The output of A/D

20 converters **734** can be connected to a neural spike detector or other suitable hardware and/or software configuration, such as a computer system, for further processing.

Referring to Figure 9, a schematic view of a signal receiver, generally designated **900**, having 32 channels and communication with a remote

computer system, generally designated **902**, according to another embodiment of the present invention is illustrated. Signal receiver **900** can include a headstage component **904** attached to the outside of the skin of the head of a subject for receiving and conditioning neural signals from the subject. Further, 5 signal receiver **900** can include a wearable component **906** for attachment to the subject and for further processing neural signal, converting the signal to a digital representation, and transmitting the signal to remote computer system **902**.

Headstage component **904** can be attached to a first group of neural 10 sensors (NS1 – NS16) **908** and a second group of neural sensors (NS17 – NS32) **910** for receiving electrical-based neural signals from the subject. The first and second group of neural signals **908** and **910** are positioned in generally different portions of the neural tissue of the subject. Headstage component **904** can also be connected to a first reference neural signal (REF 15 NS1) **912** and a second reference neural signal (REF NS2) **914** positioned generally in the same portion of the neural tissue as first and second group of neural sensors **908** and **910**, respectively, for providing a reference signal for eliminating common mode noise.

Headstage component **904** can include a first group of preamplifiers 20 (PREAMP1 – PREAMP16) **916** and a second group of preamplifiers (PREAMP18 – PREAMP33) **918** attached to first and second group of neural sensors **908** and **910**, respectively, for filtering and amplifying the neural signals. Headstage component **906** can further include preamplifiers **920** and **922** attached to reference neural sensors **912** and **914**, respectively, for filtering

and amplifying the reference neural signals. The preamplifiers can include a circuit for removing the DC offset. If the DC offset correction is not included, the output signals can connect to an analog processor of the type shown in, for example, Figure 5 which corrects the DC offset using capacitor C1 **500** and
5 resistor R1 **502** (shown in Figure 5).

Headstage component **904** can further include a first group of operational amplifiers **920** and a second group of operational amplifiers **922** for providing a difference signal of the neural signals and an associated reference neural signal to eliminate common mode noise. If operational amplifiers **922**
10 are included, the outputs of preamplifiers **916** and **918** can be connected to one of the noninverting inputs of an associated operational amplifier of first group of operational amplifiers **920** and second group of operation amplifiers **922**, respectively. The output of preamplifiers **924** and **926** can be connected to the inverting input of each of first group of operational amplifiers **920** and
15 second group of operation amplifiers **922**, respectively. Therefore, the resulting output of operational amplifiers **928** and **930** is the difference signal of the inputs, the neural signal and its associated reference signal. Operational amplifiers **928** and **930** can also provide a gain to the difference signal. Preferably, operational amplifiers **928** and **930** provide a gain of 10. If
20 operational amplifiers **922** are not included, then the signal outputs from **916**, **918**, **924** and **926** can be input to an analog processor of the type shown, for example in Figure 5.

Wearable component **906** includes an analog processor **932** for receiving and processing the signals from operational amplifiers **928** and **930**.

Analog processor **932** can condition and time-division multiplexes the neural signals. Wearable component **906** can also include an analog-to-digital (A/D) converter **934** for converting the signal into a digital representation of the neural signal for receipt by a single board computer **936**. A/D converter **934** can

5 digitize signals at 30k samples/second per channel at eight, ten, or 12 bits of resolution. Fifty digital inputs can be used for controlling channel enables, variable gain amplifiers, and analog to digital conversion timing. A/D converter **934** can include a regulated power supply **938** and pack the digitized data. In this embodiment, power supply **938** is powered by a pair of rechargeable

10 lithium-ion batteries. Computer **936** can format the data from the A/D converter into a format which is suitable for transmission over standard media for instance TCP/IP or UDP protocols for a wired or wireless network. Computer **936** can also format the data for a wired or wireless non-standard media connection such as a point-to-point wired or wireless connection. Further,

15 computer **936** can perform data reduction schemes including spike based data reduction such as spike detection and transmission or spike sorting. Additionally, computer **936** can also perform other types of data compression encoding familiar to those skilled in the art such as run length encoding or MPEG type compression. A/D converter module **934** includes power regulation

20 circuitry, complex programmable logic device (CPLD), and a first-in first-out (FIFO) memory. CPLD can generate both static and timing control signals. The static signals are specified by an operator (via the wearable computer) and loaded into registers in the CPLD. The registers automatically assert signals

onto the A/D converter and receiver modules thus controlling the function of these components.

Computer **936** can also prepare the digital signal for transmission by a wireless telemetry module **940** to remote computer system **902**. Wireless
5 telemetry module **938** and antenna **940** can transmit a representation of the neural signal to remote computer system **902** for further processing. In this embodiment, wireless telemetry module **938** comprises an IEEE (Institute of Electrical and Electronics Engineers) 802.11b wireless ethernet card for transmitting a distance up to 9 meters. Alternatively, computer **936** and
10 wireless telemetry unit **940** can be another suitable wireless technology such as 802.11a, 802.11g, a Bluetooth module, UWB radio or a commercial or other point-to-point wireless connection.

Remote computer system **902** can include a processor, memory, and a telemetry component for receiving and processing the transmitted signal from
15 antenna **942**. Computer system **902** can further process the signal and display the received neural signal and any results of processing the neural signal to an operator on a display **944**. Computer system **902** can be a commercial system for processing and receiving neural signals such as the Plexon MAP processor (plexon is in Texas) or could be a commercial computer with software to
20 process the neural signals. In either case, the signals can be spike sorted, and processed as previously described within and across signals to derive control commands from the neural signals.

Referring to Figure 10, a schematic view of a signal receiver, generally designated **1000**, having 16 channels and communication with a remote

computer system, generally designated **1002**, according to another embodiment of the present invention is illustrated. In this embodiment, signal receiver **1000** is a headstage neurochip for implantation into a subject. Signal receiver **1000** can comprise a neural signal receiver **1004** and a neural signal processor and another implanted module **1006**. Further, signal receiver **1000** can include a wearable component **1008** for attachment to the subject and for further processing neural signals, converting the signals to a digital representation, compressing the signals, and transmitting the signal to remote computer system **1002**. Further, implanted module **1006** can communicate with wearable component **1008** using radio communication or optical communication. Wearable component **1008** can relay the control data from remote computer system **1002** or use software to determine the control settings. Wearable component **1008** can also relay a portion or all of the data from the implanted component to the remote computer system **1002**. If only a portion of the data is relayed, this portion can be a compressed version of the signal or a processed version where the processing can be similar to that described above. The radio communication can be UWB radio, a commercial radio format such as 802.11a,b,g or other point-to-point radio technology. The optical communication can be near infrared light transmitted transcutaneously through the skin. The transmission can be analog or digital M-ary signals including binary. The transmission can be at other light wavelengths which are efficiently transmitted through the skin. Further, the transmission through the skin could be performed using acoustical energy in the form of analog or digital M-ary signals including binary. Such acoustical communication can be

performed with a carrier frequency of between 1 and 50 Mhz or alternately using acoustical pulses of short duration encoding the information using pulse width, pulse position or pulse amplitude modulation.

Signal receiver **1004** can comprise preamplifiers (PREAMP1 –
5 PREAMP16) **1010** attached to neural sensors (NS1 – NS16) **1012** for receiving electrical-based neural signals from the subject. Preamplifiers **1010** can filter and amplify the neural signals detected by neural sensors **1012**.

Signal receiver **1004** can also comprise a reference signal selection
component **1014** attached to preamplifiers **1010**. Component **1014** can
10 comprise multiplexers and/or differential amplifiers for selecting an appropriate reference signal. Signal receiver **1004** can also comprise a high pass filter **1016** and low pass filter **1018** for providing filtering to the neural signals. Filters **1016** and **1018** can pass all of the neural signal components (20-8000 Hz) or can pass a portion of the signal that allows for optimal spike detection and
15 sorting.

Implanted module **1006** can include an analog processor **1020** for receiving and processing the signals from signal receiver **1004**. Implanted module **1006** can also include an analog-to-digital (A/D) converter **1022** for converting the signal into a digital representation of the neural signal for receipt
20 by a processor **1024**. Processor **1024** can further process the digital signal by detecting or sorting spikes, processing spikes on a single channel by binning or some other integrating process, and combining information across neural signals to generate control signals. Processor **1024** can also prepare the digital signal for transmission as described herein by a wireless telemetry

module **1026** to wearable component **1008**. Wireless telemetry module **1026** can transmit a representation of the neural signal to remote computer system **1002** for further processing.

Wearable component **1008** can be worn directly over implanted
5 component **1000** as would be suitable for optical or acoustic transmission. Additionally, wearable component **1008** can be worn directly over or nearby if radio transmission is used. According to one embodiment, wearable component **1008** can be positioned in a helmet worn over implanted module **1006**. The components of implanted module **1006** can also be wired
10 subcutaneously such that a transmitter can reside in another region of the body and the wearable component **1008** can be positioned over the transmitter in that region. According to one embodiment, several neural signal receiver modules can be placed near the electrodes and the transmitter placed subcutaneously near the pectoral region of the chest. These modules can be
15 connected together using subcutaneous wires tunneled inside the body. The wearable relay system can be placed in a vest or jacket with telemetry module **1026** positioned over the implanted transmitter module.

Remote computer system **1002** can include a processor, memory, and a telemetry component for receiving and processing the transmitted signal from
20 antenna **1028**. Computer system **1002** can further process the signal as described herein and display the received neural signal and any results of processing the neural signal to an operator on a display **1030**. Computer system **1002** can also process the signals as described above to derive control signals to control a prosthetic device or other devices as described above.

Referring to Figure 11, a schematic view of a neurochip **1100** including a wireless telemetry module **1102** implanted subcutaneously in communication with a wearable component **1102** positioned outside the skin and adjacent neurochip **1100**. Wireless telemetry module **1102** can comprise a neural signal
5 transmitter **1106** for transmitting conditioned neural signals. Wearable component **1104** can comprise a neural signal receiver **1108** for receiving the conditioned neural signals.

Wireless telemetry module **1102** and wearable component **1104** can communicate control signals to a control module, such as control module **436**
10 shown in Figures 4A and 4B, for controlling various components of signal conditioner module, such as signal conditioner module **402** shown in Figures 4A and 4B, and an A/D converter module, such as converter module **404** shown in Figures 4A and 4B. Wearable component **1104** can transmit control signals to neurochip **1100** via a transmitter **1110**. The control signals can
15 originate from a processor in wearable component **1104** running an algorithm as described below, or can relay the control information from a remote processor under operator control or under the control of an algorithm. Wearable component **1104** can also receive data signals from neurochip **1100** via a transmitter **1112** and process or compress the data signals. Wearable
20 component **1104** can also transmit the modified data to a remote receiver for further processing and for device control.

The control signals can be generated by a processor that examines the neural signals to determine the usefulness of a given signal. If a given signal is considered useful, for example because it contains characteristic spikes as

determined by an operator or an algorithm, then the control signals which control the power to that channel are enabled and the control signals which control the multiplexer and A/D converter are enabled to sample and convert the given channel. Also, if a channel is enabled, the control signals controlling the amplification and filtering are manipulated to provide an optimized neural signal for later detection and sorting. For instance, the gain can be adjusted by an operator or an algorithm such that the amplitude of the largest spikes fill 2/3 of the A/D converter input range. Similarly, the filter setting can be optimized by an operator or an algorithm to give a characteristic neural waveshape. The control signals can be generated by an analog or digital processor that is located in neurochip **1100**, located in the implanted signal processor, one of the wearable components or in the remote signal receiver. They could originate from a processor performing an algorithm as described, or could be controlled through the processor by an operator. Wireless telemetry module **1102** can comprise a control receiver **1114** for receiving the control signals. Wearable component **1104** can comprise a control transmitter **1116** for transmitting control signals to control receiver **1114**.

Wearable component **1104** can subcutaneously transmit power to neurochip **1102**. Wearable component **1104** can comprise a power/clock transmitter for transmitting power and a clock signal electromagnetically to wireless telemetry module **1102**. Wireless telemetry module **1102** can comprise a power/clock receiver **1120** for receiving the power and clock signal from power/clock transmitter **1118**. The clock signal can be important when there is more than one neural signal receiver, as spikes obtained from different

anatomical locations need to be synchronized to a single clock source. The clock source can be generated by a processor or IC in wearable component **1104** or in a remote receiver.

5

IV.B. Neural Spike Detection

Referring to Figure 12, a schematic view of a neural spike detector, generally designated **1200**, according to one embodiment of the present invention is illustrated. Neural spike detector **1200** can include an analog sampler **1202** for receiving the analog conditioned neural signals of a single
10 neural sensor from a signal receiver. Analog sampler **1202** samples the conditioned signal N times and stores the N samples. Neural spike detector **1200** can include a counter and control logic module **1204**, decoder **1206**, and control logic/memory module **1208** for controlling sampling. Module **1204** can be operable to receive a clock signal input **1210** for synchronization with a
15 signal receiver or neural spike processor. In one embodiment, the N samples are stored in analog form such as by a low leakage capacitor. Alternatively, the N samples can be stored in digital form. Neural spike detector **1200** can include a buffer array **1212** having N buffers for buffering and driving each of the N samples. Neural spike detector **1200** can also include an analog
20 processor **1214** for receiving the N samples from buffer array **1212**. In one embodiment, analog processor **1214** combines the N samples by implementing a scaling (gain) and summing algorithm. Alternatively, the N samples can be combined using gain and/or power operations known to those of skill in the art. Analog processor **1214** characterizes the results of the operations using a

comparison to threshold levels in the processor and output to a logic state machine **1216**. Logic state machine **1216** determines whether a neural spike has been detected. On the detection of a spike, the N samples in buffer array **1212** can be passed to an output spike formatter **1218** for transmission for
5 further processing, such as spike sorting and interpretation. The output of output spike formatter **1218** is the N samples stored in buffer array **1212** and M samples that are immediately following the N samples. The M samples can be used for further characterizing the neural spike. In one embodiment, the sampling rate is 40 kHz, N is 8, and M is 32. Control logic/memory module
10 **1108** can control operation of analog processor **1214** and logic state machine **1216**.

Referring to Figure 13, a schematic view of a neural spike detector, generally designated **1300**, according to one embodiment of the present invention is illustrated. Neural spike detector **1300** can include an analog-to-
15 digital converter **1302** for receiving the analog conditioned neural signals of a single neural sensor from a signal receiver. Analog-to-digital converter **1302** can convert the conditioned neural signal into a digital representation. Neural signal detector **1302** can also include a digital processor **1304** for detecting neural spikes in the neural signal. In one embodiment, neural spikes are
20 detected by (1) sampling the conditioned signal; (2) combining the samples; (3) comparing the combined samples to predetermined thresholds; and (4) determining whether a spike has been detected based on the comparison in step (3). Digital processor **1304** can transmit samples of a neural spike to an output spike formatter **1306** for transmission to a device for further processing.

IV.C. Neural Spike Processor and Transmitter

Referring to Figure 14, a schematic view of a signal transmitter, generally designated **1400**, according to an embodiment of the present invention is illustrated. Signal transmitter **1400** is advantageous because signal
5 transmission includes UWB transmission, wherein data is transmitted only when a neuron spike is detected on any of the input neural signals from neural sensors (NS1 – NS16) **1402**. UWB transmission saves significant power consumption. Signal transmitter **1400** includes an asynchronous wideband protocol having channel identifiers and time stamp coding so that when a spike
10 is detected on any one channel the time and source can be uniquely identified.

Signal transmitter **1400** can include signal conditioners **1404** and neural spike detectors **1406**. Each neural sensor **1402** can be connected to one of signal conditioners **1404** for providing filtering and amplification to the detected neural signals. Signal conditioners **1404** can be connected to neural spike
15 detectors **1406**. Neural spike detectors **1406** are operable to detect a neural spike on the conditioned neural signal and transmit a pulse signal on the detection of a neural spike. Neural spike detectors **1406** can generate a UWB pulse sequence consisting of pulses of approximately one nanosecond in duration on the detection of a neural spike.

20 Generally spike signal trains have a repetition rate of between approximately 10 and 50 Hertz and spike duration of between approximately 1 and 2 milliseconds. Thus, there is significant dead time between the neural spikes allowing all of the spike information about one spike to be transmitted before another spike on the same channel occurs. The wide band nature of

signal transmitter **1400** is generated from the very narrow (~1 nanosecond) digital pulse signals that are generated and transmitted to an antenna driver **1408** and antenna **1410**.

Signal transmitter **1400** can include a collision detector and separator
5 component **1412** for preventing any overlap of neural spike pulses occurring simultaneously from all of neural spike detectors **1406**. On the detection of an overlap, component **1412** queues the pulses and transmit the pulses to channel identifiers **1414**. Channel identifiers **1414** can replace each pulse with a generated channel identification code having m bits for indicating the origin
10 neural sensor **1402**. The channel identifier encoder that creates a unique digital code (m bits in length) for each channel is provided by the equation $2^m = n$.

The digital data generated by component **1412** is summed together and transmitted to a sync pulse generator **1416**. Generator **1416** interleaves a
15 period sync pulse with the data pulses. The sync pulse can keep the signal receiver, described below, and signal transmitter **1400** in synchronization. The sync and data pulses can then be convolved with the signal generated by a VCO **1418**. The resulting signal is transmitted to antenna driver **1408** and antenna **1410** for transmission. The pulsewidth of the baseband signal
20 determines how many VCO cycles will be transmitted per symbol. In this embodiment, each sync and data pulse is represented by half cycles of the VCO waveform, but can also be increased by integer multiples for improved symbol recover rate in the presence of noise within the signal receiver.

Referring to Figure 15, a schematic view of a signal receiver, generally designated **1500**, and a spike processor, generally designated **1502**, according to one embodiment of the present invention is illustrated. Signal receiver **1500** can receive signal from signal transmitter **1400** of Figure 14. Signal receiver

5 **1500** can include a linear amplifier matching filter **1502** for receiving the transmitted signal. The received signal is passed to a sync waveform correlator and peak detector **1504** for recovering the delay/timing of each transmitted frame resulting in a pulsed waveform. The pulsed waveform is transmitted to channel data correlators **1506** for identifying the time when the data is

10 recovered. Next, channel spike recovery components **1508** convert the channel-encoded signals to the actual single pulse output. The individual outputs are then sent to a spike processor **1510** for processing into control signals.

It will be understood that various details of the invention may be

15 changed without departing from the scope of the invention. Furthermore, the foregoing description is for the purpose of illustration only, and not for the purpose of limitation.